A New Analytical Model for Improving Subthreshold Conduction of MOSFETs

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The sub-threshold current conduction in MOSFETs has already been discussed in a great detail in literature. Improvement of sub-threshold current slope of MOS devices still remains a concern due to static power dissipation in dynamic circuits and standby CMOS power consumption. Here, we propose a long channel MOSFET structure with larger effective channel thickness and a mathematical model to represent it, which reduces the magnitude of sub-threshold current to pico ampere level. However, inclusion of oxide charges, mobile ions and interfacial traps increase it to nano ampere range.

The diffusion current through the channel in subthreshold region of operation can be derived into the following equation [1],

$$I_D = \frac{w}{L} x_d \mu_n k T n_{po} e^{\beta \psi_s} (1 - e^{-\beta v_D})$$
(1)

Where, β is equal to $\frac{kT}{q}$, v_D is the drain voltage, ψ_s is the surface potential, x_d is the effective channel thickness and all other parameters bear the usual meaning found in the literature.

To find out the relation between potential and depletion depth, at first 1-D Poisson equation is solved at the surface charge region. The solution gives potential as a function of the distance along the depth of the depletion region (fig.1). A tangent is drawn in fig. 1 at $x=0, \psi = \psi_s$. It is assumed that the intersection of the tangent with x axis is the effective channel thickness which is larger than the previous assumption in [1]. Substituting the new effective channel thickness in (1), current is found to be,

$$I_{D} = \frac{w}{L} \mu_{n} k T n_{po} e^{\beta \psi_{s}} (1 - e^{-\beta v_{D}}) \psi_{s} \varepsilon_{s} \cdot \left[2 \left(\frac{\psi_{s}}{\beta} + e^{\beta \psi_{s}} + j \right) (1 + e^{\beta \psi_{s}}) \beta e^{\beta \psi_{s}} - (1 + e^{\beta \psi_{s}})^{2} \left(\frac{1}{\beta} + \beta e^{\beta \psi_{s}} \right) \right] \\ \frac{b(1 + e^{\beta \psi_{s}})^{4} \beta q}{b(1 + e^{\beta \psi_{s}})^{4} \beta q}$$
(2)

Where,
$$\frac{p_{po}e^{-\beta\psi_s} - n_{po}e^{\beta\psi_s}}{\left(n_{po} - p_{po}\right)} - \beta\psi_s = j$$

and
$$\sqrt{\frac{\varepsilon_s}{2q}} \frac{2}{\sqrt{\beta}} \frac{1}{(1+e^{\beta\psi_s})} \sqrt{\frac{\psi_s}{\beta} + e^{\beta\psi_s} + j} = b$$

The resultant sub-threshold current is shown in fig. 2. Effects of fixed oxide charge, mobile ion [2] and interfacial traps [3] are included in (2) because $V_{GB}=\psi_s+\psi_{ox}+\phi_{MS}$. V_{GB} is replaced by $V'=V_{GB}+\Delta V_{GB}$, where ΔV_{GB} accounts for changes in gate voltage due to non-ideal effects [2]. Fig. 3 shows the comparison between sub-threshold conduction in ideal and non-ideal MOSFET. Changes in V_{GB} due to oxide charges, mobile ions and interfacial traps are calculated from corresponding formulas[2-3].



Fig. 3.Sub-threshold conduction at ideal and non-ideal condition vs. drain voltage

References

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